

Studienarbeit: Planar Multi-Beam Antenna for W-LAN

Presented by
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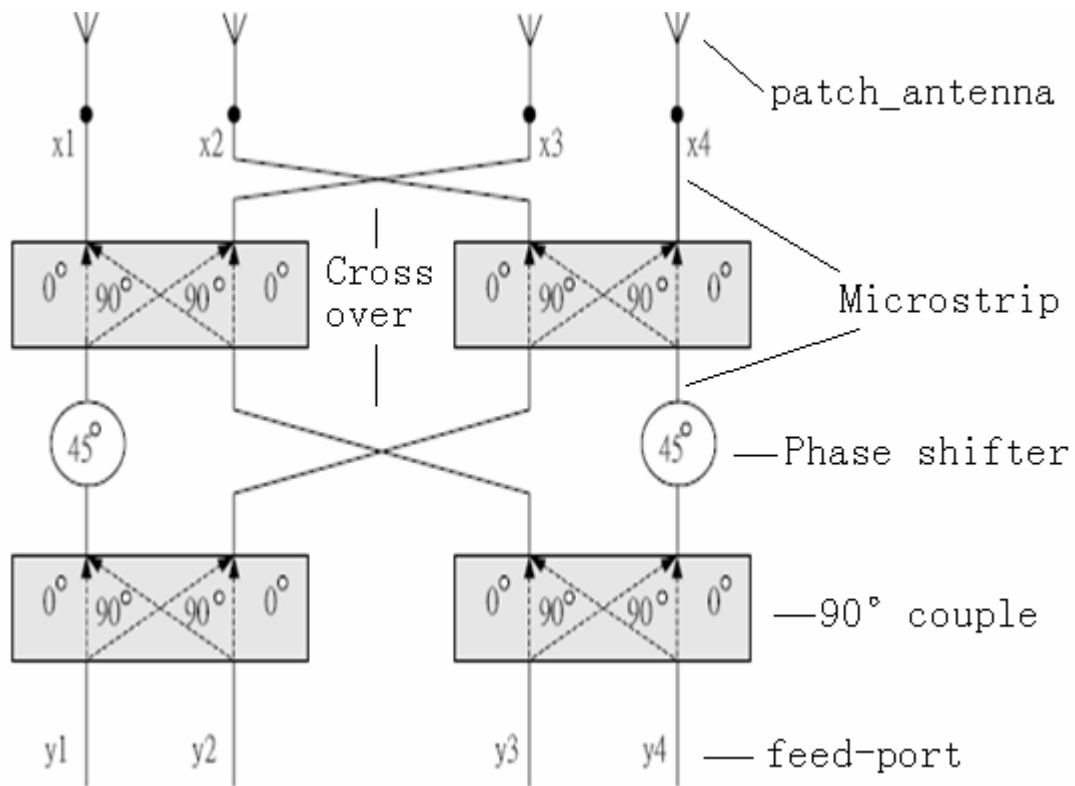
Duisburg
19. 11. 2007

1. Introduction

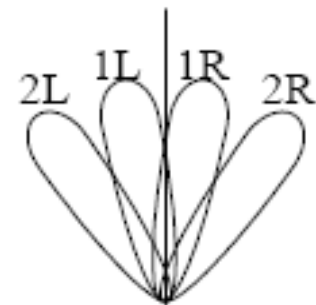
The switched-Beam antenna is one type of the small antennas, which consists of the antenna array and the beam forming network. The four-beam smart antenna generates four beams to cover 120° area.

In this thesis, we design the four-beam forming network called **Butler Matrix** (at frequency 2.4 GHz), which made of four directional coupler-power dividers (90° phase shifter) and two cross coupler (0dB). And we use microstrip antennas as array elements.

For a special requirement, we can reduce the element spacing and bend the planar substrate carrying the radiator elements, to get coverage of a full 180° angular in azimuth (hemisphere).



4x4 Butler Matrix

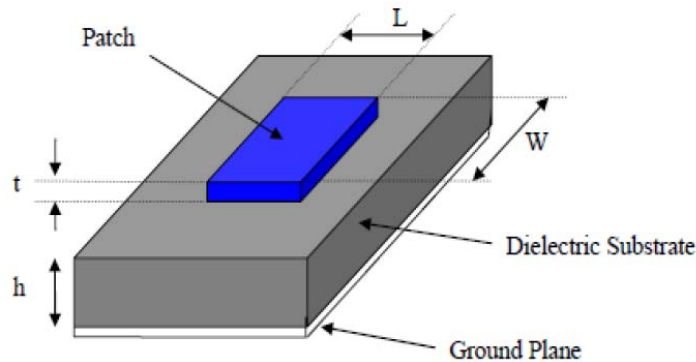


	x1	x2	x3	x4	Main beam
y1	-45	-90	-135	-180	1R
y2	-135	0	-225	-90	2L
y3	-90	-225	0	-135	2R
y4	-180	-135	-90	-45	1L

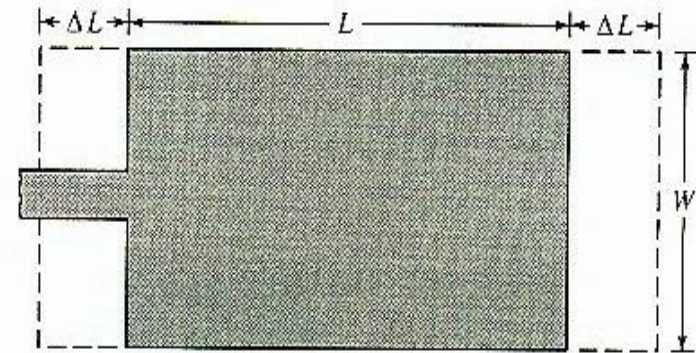
(degree)

2. Patch Antenna

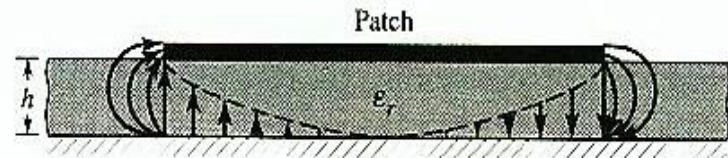
- 2.1 Transmission-Line Model



Basic structure of a microstrip antenna



(a) Top view



(b) Side view

Fringing Effekt

$$L_{\text{eff}} = L + 2 \Delta L$$

L : physical Length

L_{eff} : effektive Length

$$\epsilon_{\text{reff}} = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \left[1 + 12 \frac{h}{W} \right]^{-1/2}$$

$$\frac{\Delta L}{h} = 0.412 \frac{(\epsilon_{\text{reff}} + 0.3) \left(\frac{W}{h} + 0.264 \right)}{(\epsilon_{\text{reff}} - 0.258) \left(\frac{W}{h} + 0.8 \right)}$$

$$L = L_{\text{eff}} - 2 \Delta L$$

$$L_{\text{eff}} = \frac{\lambda_{\text{eff}}}{2} = \frac{1}{2} \frac{C_0}{fr \sqrt{\epsilon_{\text{reff}}}}$$

h: Thickness of Substrate W: Width of Patch fr: Resonant Frequency

2.2 Antenna Design procedure

a) The substrate we have is ROGER4003, with

$$\epsilon_r=3.38, h=0.51\text{mm}(35\mu\text{m Cooper}), \tan\delta=0.0027$$

b) For an efficient radiator, we let

$$W = \frac{v_0}{2fr} \sqrt{\frac{2}{\epsilon_r + 1}} = \frac{3 \times 10^8}{2(2.4 \times 10^9)} \sqrt{\frac{2}{3.38 + 1}} = 42.23\text{mm}$$

c) Using Transmission-Line Model, we have

$$\epsilon_{\text{eff}} = \frac{3.38 + 1}{2} + \frac{3.38 - 1}{2} \left[1 + 12 \frac{0.51}{42.23} \right]^{-1/2} = 3.30$$

$$\Delta L = 0.51(0.412) \frac{(3.30 + 0.3) \left(\frac{42.23}{0.51} + 0.264 \right)}{(3.30 - 0.258) \left(\frac{42.23}{0.51} + 0.8 \right)} = 0.243\text{mm}$$

$$L_{\text{eff}} = \frac{\lambda}{2} = \frac{1}{2} \frac{1}{2.4 \times 10^9 \sqrt{3.30}} = 34.41\text{mm}$$

$$L = 34.41 - 2 \times 0.243 = 33.92\text{mm}$$

d) Microstrip-Line inset-length

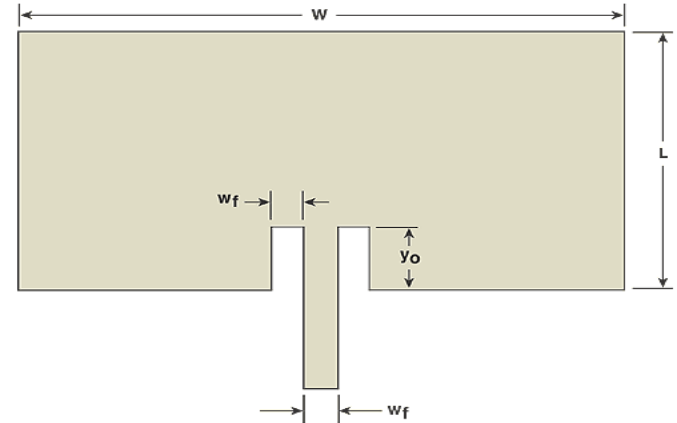
From literature, we can approximately have y_0 for characteristic

impedance 50 ohm

$$y_0 = \frac{L}{2\sqrt{\epsilon_{\text{reff}}}} = \frac{33.92}{2\sqrt{3.30}} = 9.34 \text{ mm}$$

Let $W_f = b = 1.141 \text{ mm}$

(b : microstrip line width for 50 Ohm)



Also from (<http://mwrf.com/Articles/Index.cfm?ArticleID=6993>) we can get the exact inset length for 50 Ohm input impedance through the mathematical model

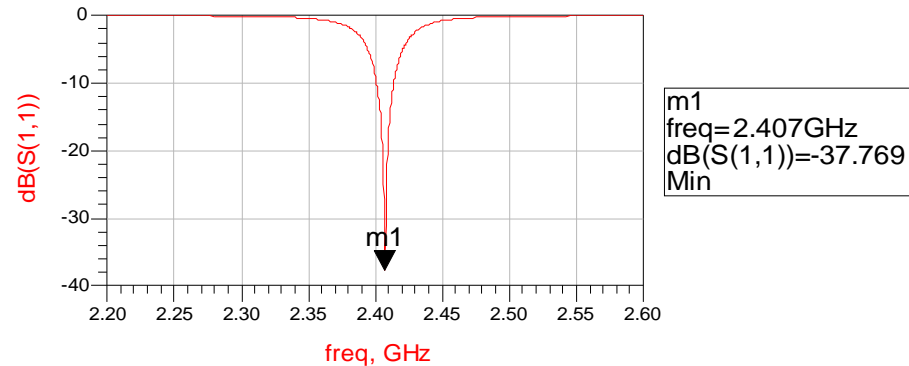
$$y_0 = 10^{-4} \left\{ 0.001699\epsilon_r^7 + 0.1376\epsilon_r^6 - 6.1783\epsilon_r^5 + 93.187\epsilon_r^4 - 682.69\epsilon_r^3 + 2561.9\epsilon_r^2 - 4043\epsilon_r + 6697 \right\} \frac{L}{2} \quad 2 \leq \epsilon_r \leq 10$$

We have $y_0 = 9.48 \text{ mm}$

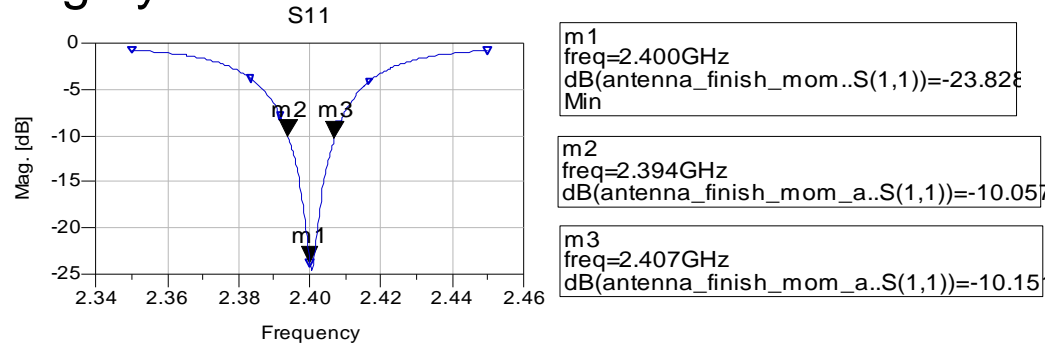
2.3 Simulation of the patch antenna

From 2.2 we have $W=42.23\text{mm}$, $L=33.92\text{mm}$, $w_f=1.141\text{mm}$, $y_0=9.48\text{mm}$ for the antenna.

Using ADS---Momentum, we draw the antenna and have the Simulation result :



Then we change the $L=34.05\text{mm}$, to get $f_r=2.4\text{GHz}$. At the same time, we must change $y_0=9.515\text{mm}$. And the simulation result is:



Finally we have $W=42.23\text{mm}$, $L=34.05\text{mm}$, $y_0=9.515\text{mm}$. And the bandwidth (at -10dB) of the antenna is approximately $B=13\text{MHz}$ (0.542%).

3. Butler-Matrix

3.1 4x4 Butler-Matrix

For Butler Matrix, from literature, we have the observation angle of the Main beam

$$\theta = \sin^{-1}\left(-\frac{\beta}{kd}\right) = \sin^{-1}\left(-\frac{\lambda\beta}{2\pi d}\right)$$

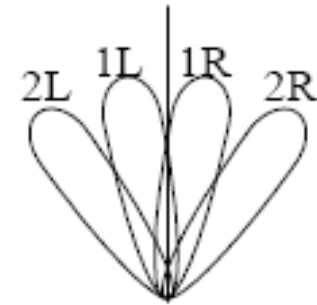
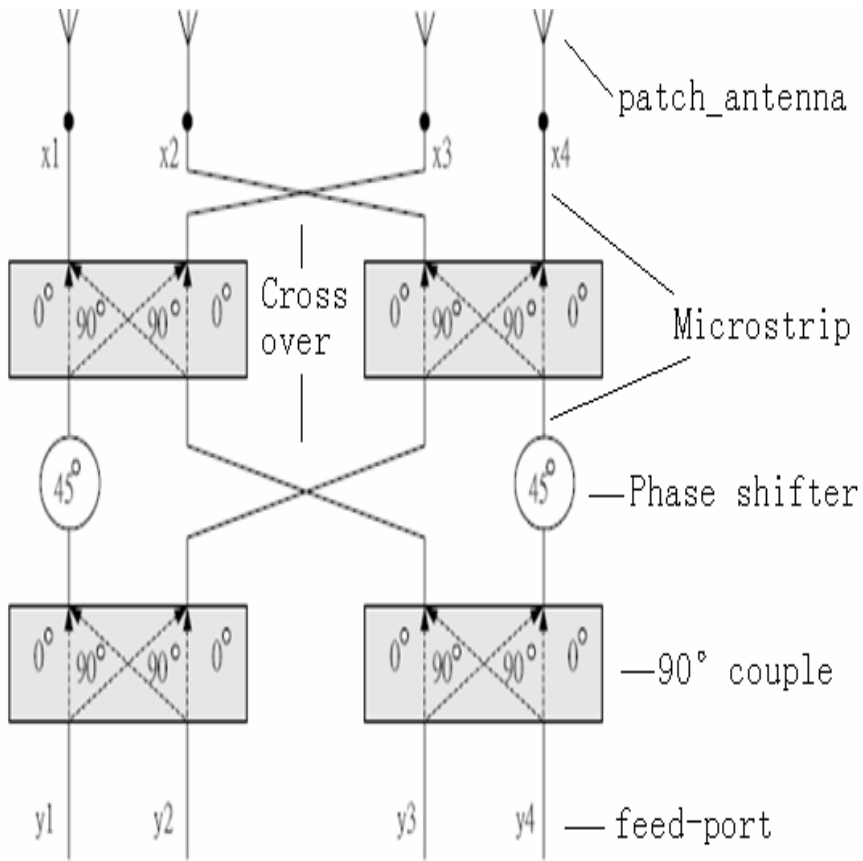
β is the difference of phase between any two successive elements, which is decided by Butler-Matrix

$$\beta = \pm \frac{(2i - 1)\pi}{N} \quad i = 1, 2, 3 \dots \frac{N}{2}$$

d is the spacing between antennas, by reducing the d , we can have bigger θ

For a 4x4 Butler-Matrix, we have $N=4$, we let $d = \frac{\lambda}{2}$

Then we can calculate the observation angle of the Main beam (θ)



	x1	x2	x3	x4	β	θ
y1	-45°	-90°	-135°	-180°	-45°	14.5° (1R)
y2	-135°	0°	-225° (135°)	-90°	135°	-48.6° (2L)
y3	-90°	-225° (135°)	0°	-135°	-135°	48.6° (2R)
y4	-180°	-145°	-90°	-45°	45°	-14.5° (1L)

4x4 Butler-Matrix

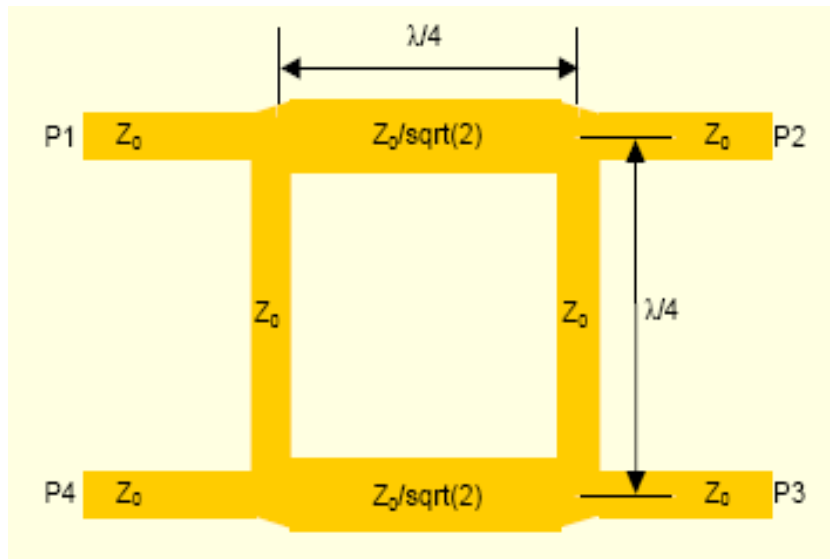
3.2 3dB 90°Hybrid Coupler and Cross Coupler

Using ADS---LineCalc

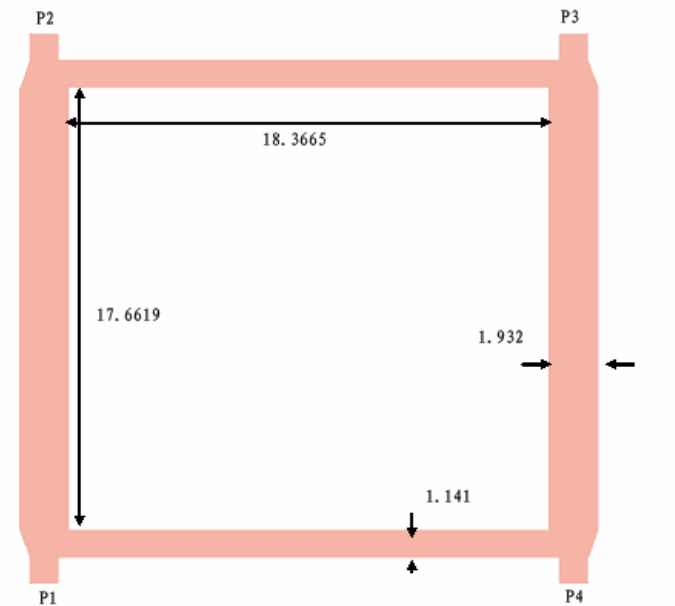
a) For $Z_0=50$ Ohm, $W=1.141$ mm

b) For $Z_0=\frac{50}{\sqrt{2}}$ Ohm, $W=1.932$ mm

3.2.1 3dB 90°Hybrid Coupler

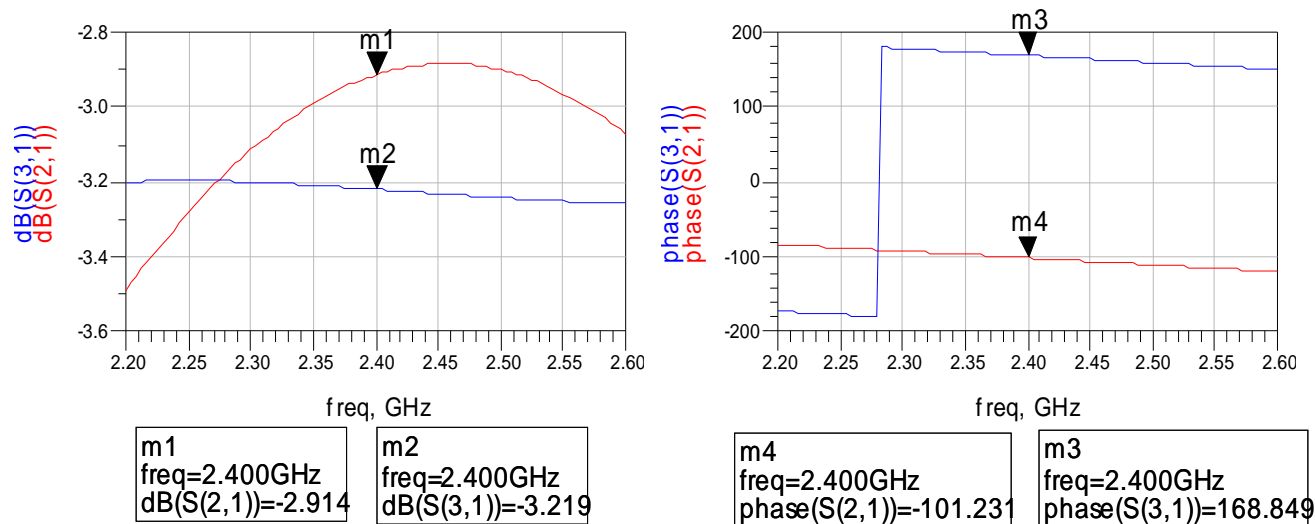


90° hybrid coupler built around transmission lines



The optimized 90°Hybrid Coupler (mm)

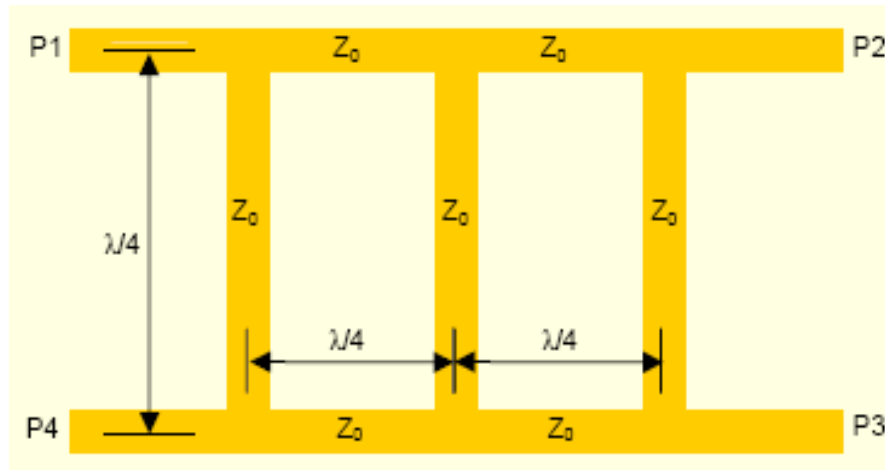
And the simulation result of the Hybrid Coupler is



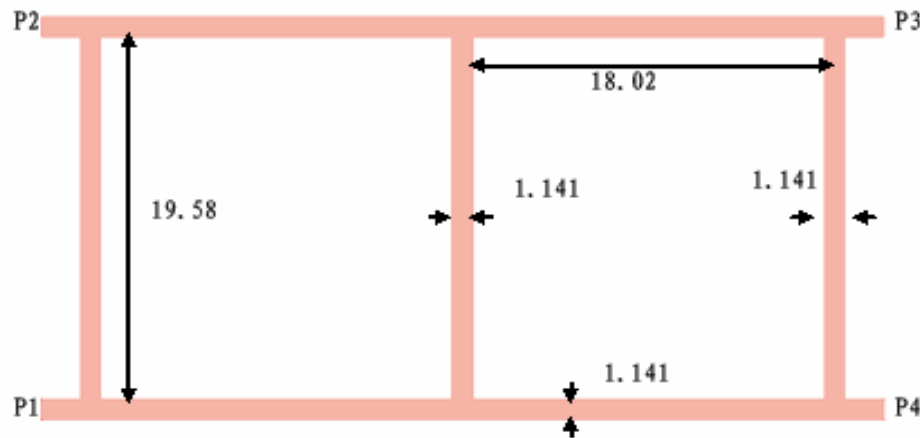
As expected, the phase difference between port 3 and port 2 is $168.849 - (-101.231) = 270.08$ (-89.92) degrees.

And the dB(S(2,1)) dB(S(3,1)) are close to -3dB.

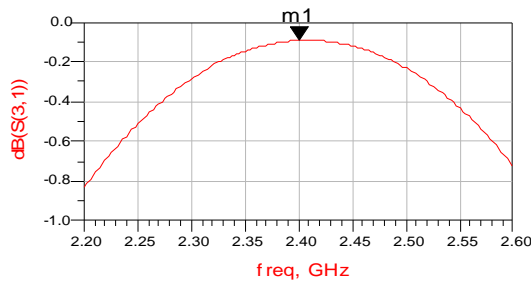
3.2.2 Cross Coupler



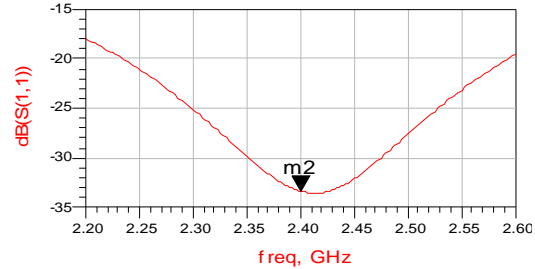
0dB Cross Coupler build around transmission lines



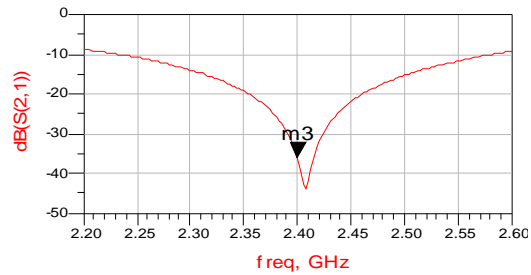
The optimized Cross Coupler (mm)



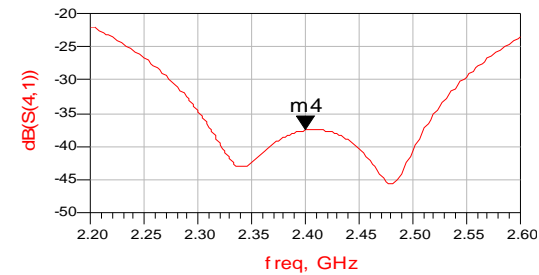
m1
freq=2.400GHz
dB(S(3,1))=-0.093



m2
freq=2.400GHz
dB(S(1,1))=-33.364



m3
freq=2.400GHz
dB(S(2,1))=-35.820



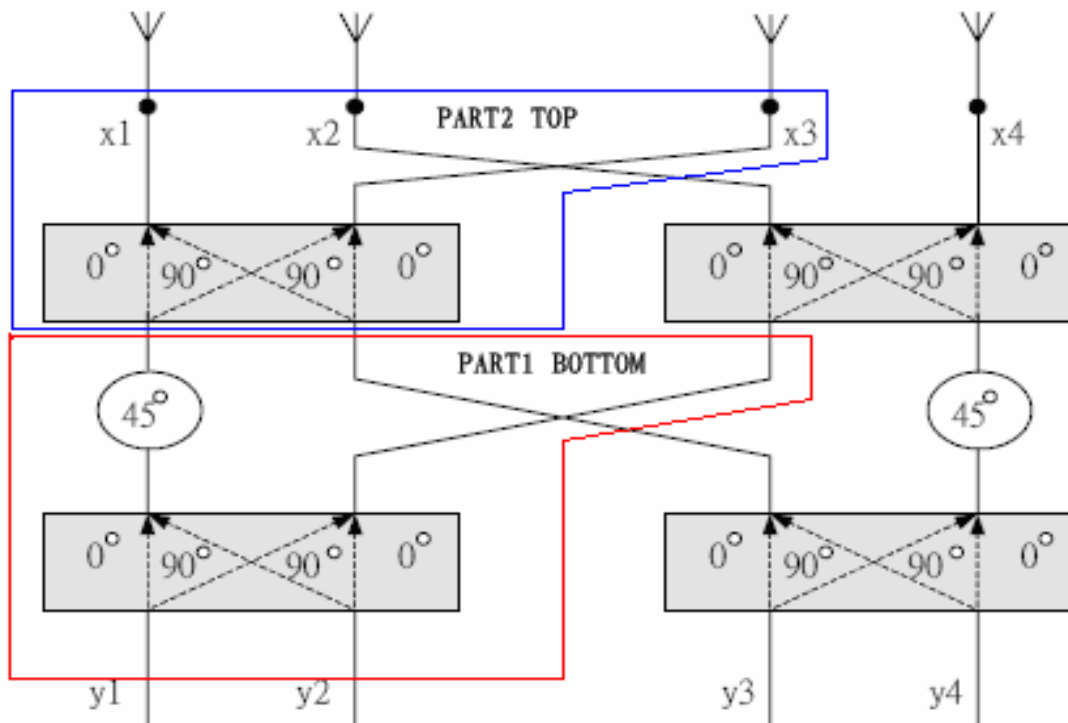
m4
freq=2.400GHz
dB(S(4,1))=-37.553

Simulation results of the Cross Coupler

The insertion loss for the coupled port is $\text{dB}(S(3,1)) = -0.093$ dB while isolation ports are smaller than -30 dB

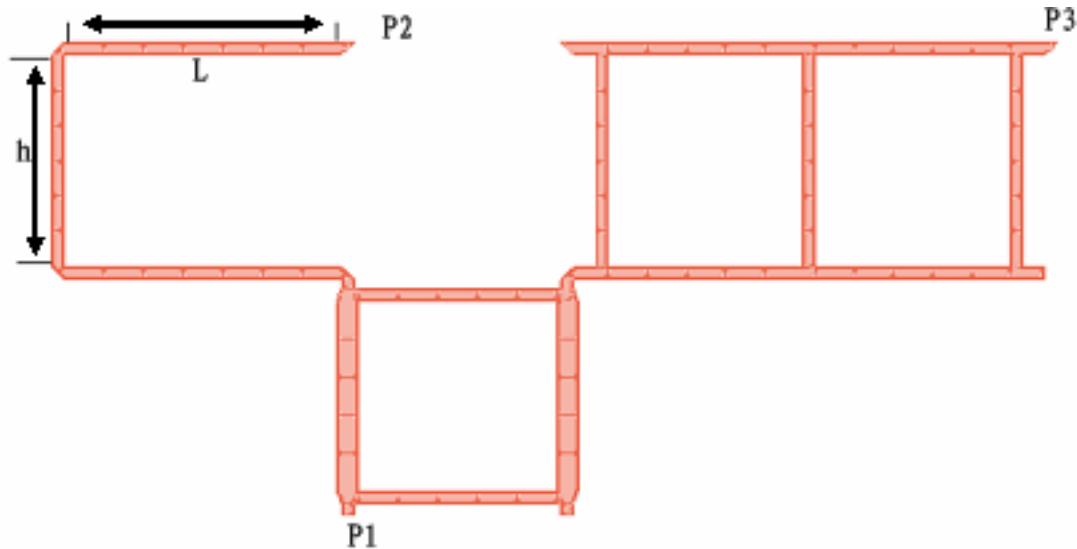
3.3 Optimization and simulation of 4x4 Butler-Matrix Network

The left and the right parts of the Butler Matrix are symmetric, so we can just consider the left side of the Butler Matrix. Now we divide the left side into 2 Parts: PART1 (BOTTOM) and PART2 (TOP).



Two Parts of the Butler Matrix (Bottom and Top)

3.3.1 Part 1 (Bottom)



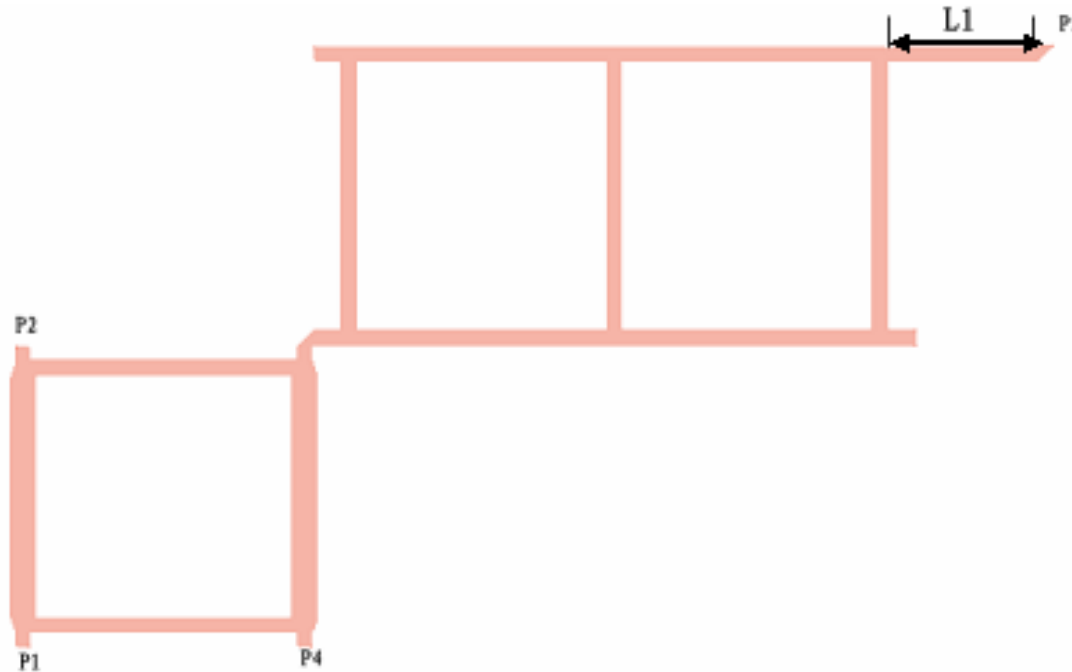
- Simulation of the right side Phase ($S(3,1)=-129.140^\circ$)
- We can calculate $h=19.4818\text{mm}$ (the same height with right)
- $S=S_{31}-S_{21}$ should be -45° , allow L to vary, set the optimization $S_{21}=-84.14^\circ$ After optimization we have $L=25.61\text{mm}$.

3.3.2 Part2: (Top)

- a) Consider right side of the Top, we must let $L1=10.8989\text{mm}$ to get the spacing between antennas

$$d = \frac{\lambda_0}{2} = \frac{1}{2} \frac{C_0}{fr} = 62.5\text{mm}$$

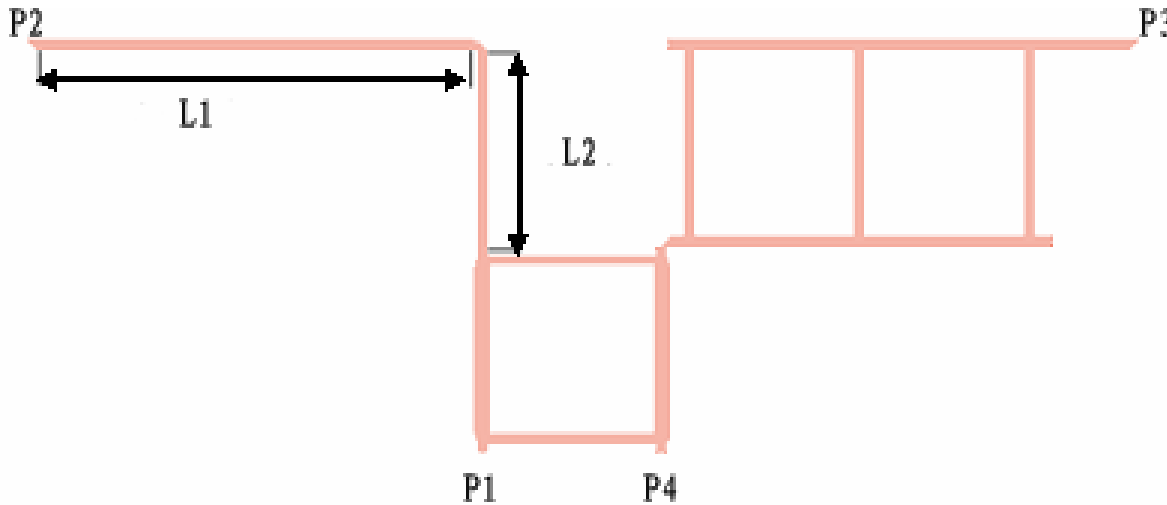
- b) Then we have simulation result $S31=-170.987^\circ$



The right side of the Top

c) Consider the left side, we can calculate $L_2=20.721\text{mm}$, we must let $L_1=49.861\text{mm}$ to get the space between antennas $d=62.5\text{mm}$

d) We have simulation result $S_{21}=-78.783^\circ$

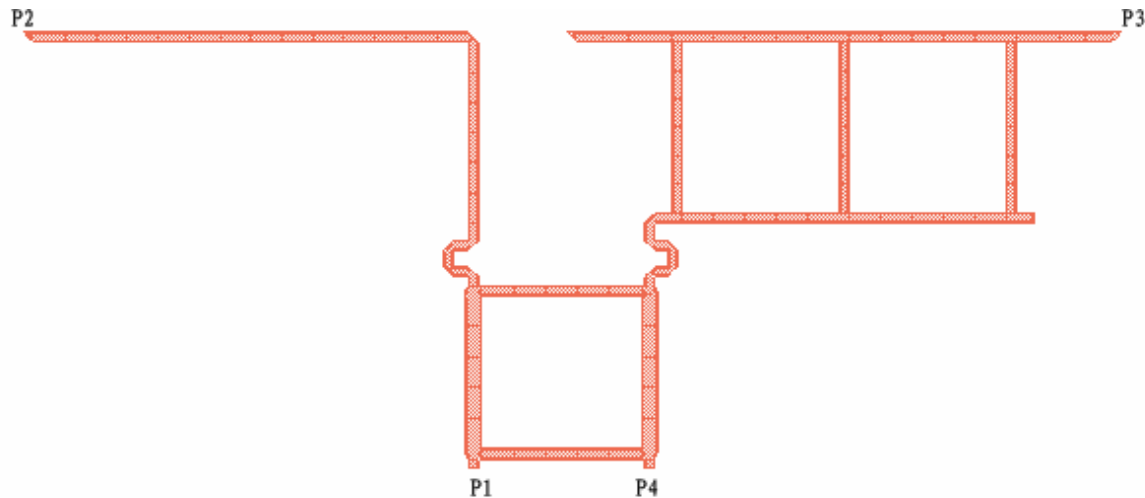


The left side of Top

e) We have $S_{31}=-170.987^\circ$ and $S_{21}=-78.783^\circ$.

Then we have $S=S_{31}-S_{21}=-92.2037^\circ$ (Target: -90)

- f) This means we should increase $S_{21} = -2.2037^\circ$,
 (also $L_{21} = 0.4708\text{mm}$). It is difficult such a short length to
 increase with bends.
- g) Then we have 2 choices:
 (1) Increase $S_{21} = -2.2037^\circ - 360^\circ = -362.2037^\circ$
 (also $L_{21} = 77.3799\text{mm}$)
 (2) Increase S_{31} , then increase $S_{21} = S_{31} + (-2.2037^\circ)$
- h) We choose (2), we first increase S_{31} (arbitrarily), we have the
 simulation result $S_{31} = 145.302^\circ$

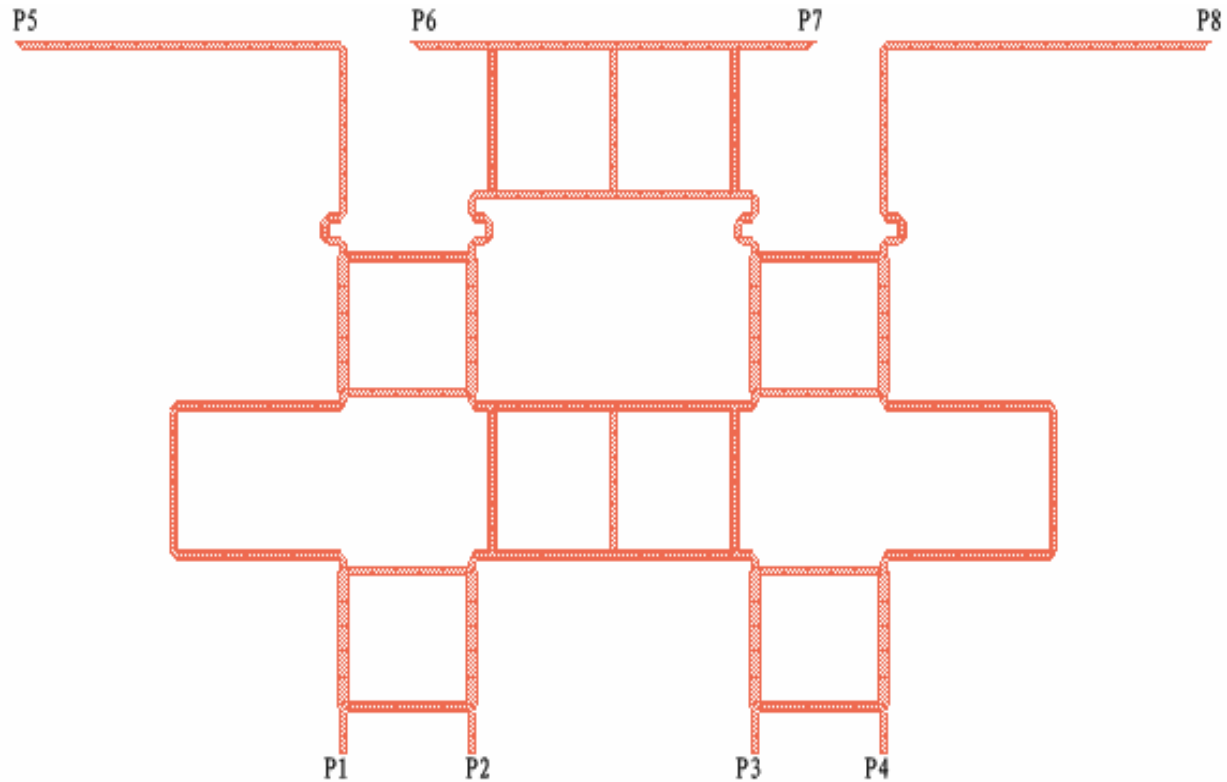


The top of the Butler-Matrix

- i) Then we setup the goal of the optimization $S_{21} = -124.698^\circ$
($S = -90^\circ$). After optimization we can confirm the design of the left side.

3.4 Simulation of the Butler Matrix Network

From 3.3 and the symmetry of the Butler Matrix, we have the Butler Matrix Network



Butler-Matrix Network

The simulation results are shown in the Table

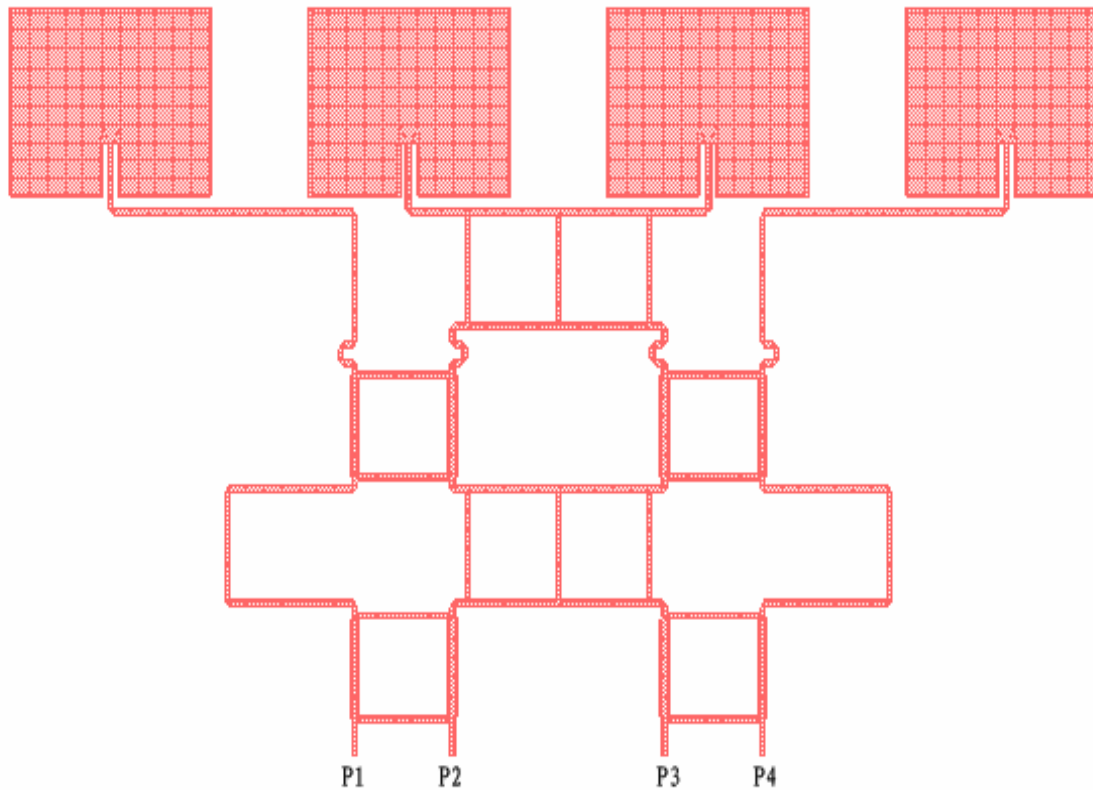
	P5	P6	P7	P8	β
P1	126.272	84.418	39.043	-6.513	-45
P2	38.422	171.941	-54.383	83.405	135
P3	83.361	-54.481	171.935	38.387	-135
P4	-6.516	38.985	84.435	126.237	45

The phase difference between input and output ($^{\circ}$)

- a) For input P1, we have Phase difference -41.854° , -45.375° , -45.556°
- b) For input P2, we have Phase difference 133.519° , -226.324° (133.676°), 137.788°
- c) For input P3, we have Phase difference -137.842° , 226.416° (-133.584°), -133.548°
- d) For input P4, we have Phase difference 45.501° , 45.45° , 41.802°

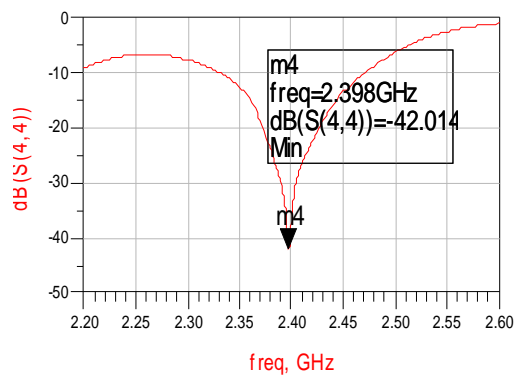
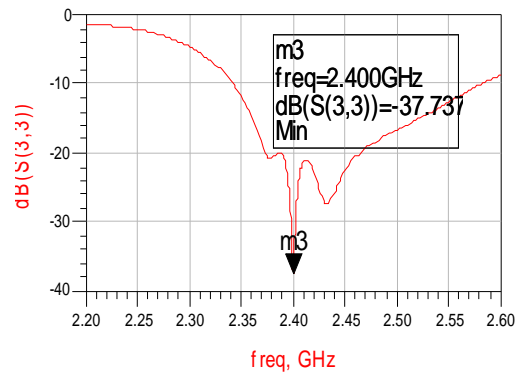
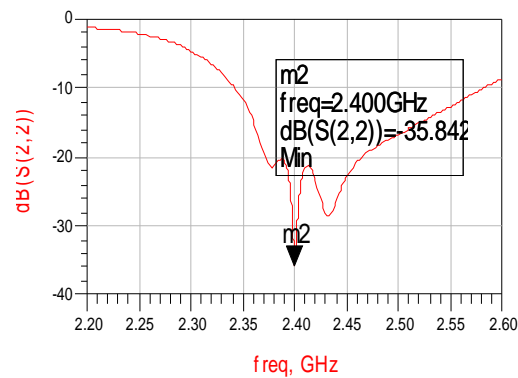
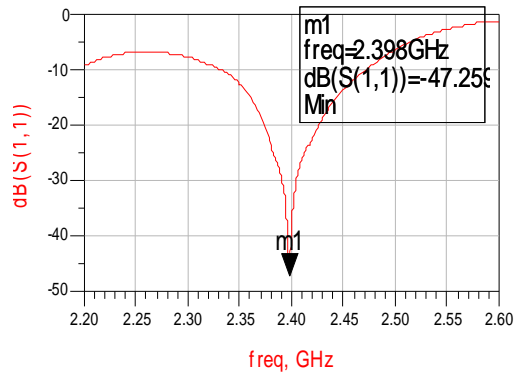
3.5 The simulation results of Butler-Matrix (with antennas)

We have already the network of the Butler Matrix Now we just add the 4 patch antennas (see 2.3)

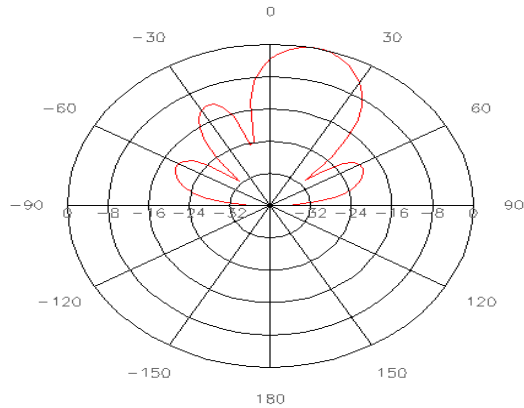


Butler Matrix with antennas

The simulation results are



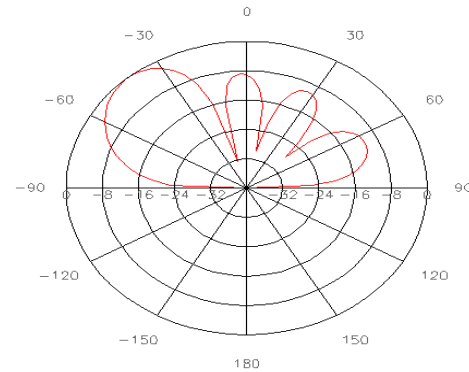
Using Momentum---Post Processing---Radiation Pattern Control, we have 2-D Polar Far Field of the Butler Matrix



Fri Nov 02 03:25:57 2007

antenna_with_network

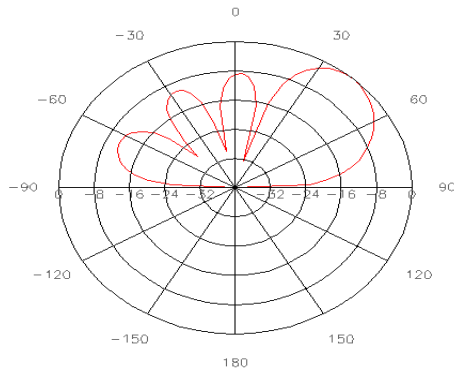
For Input P1



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antenna_with_network

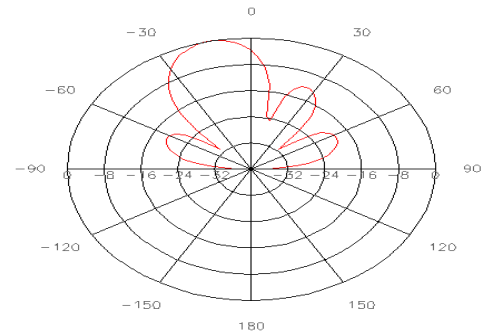
For Input P2



Fri Nov 02 03:30:31 2007

antenna_with_network

For Input P3



Fri Nov 02 03:32:59 2007

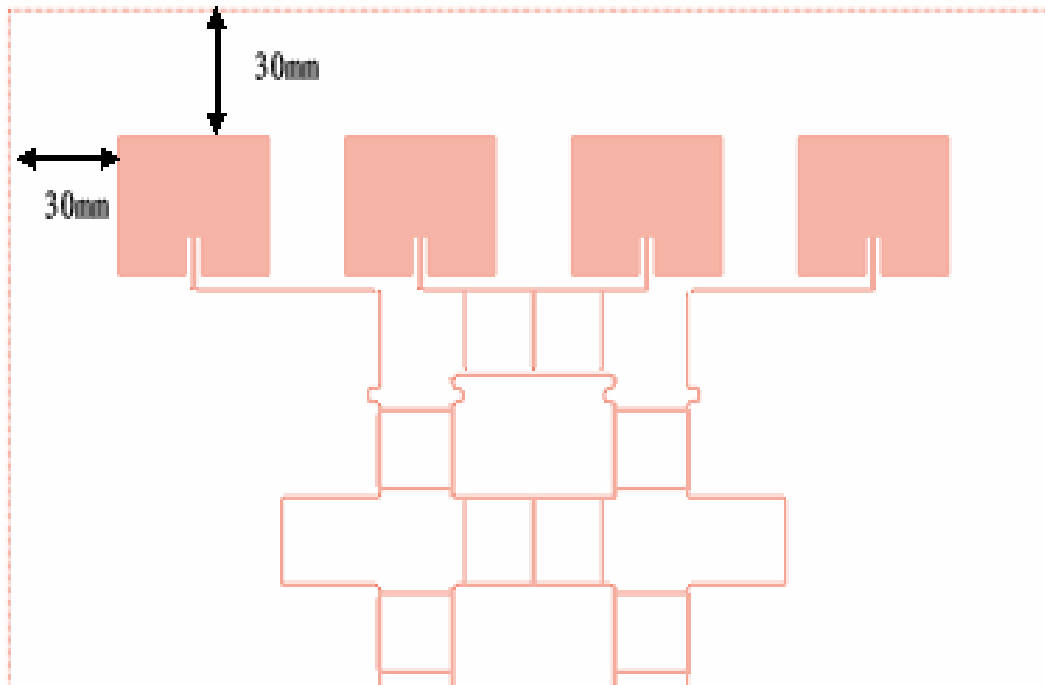
antenna_with_network

For Input P4

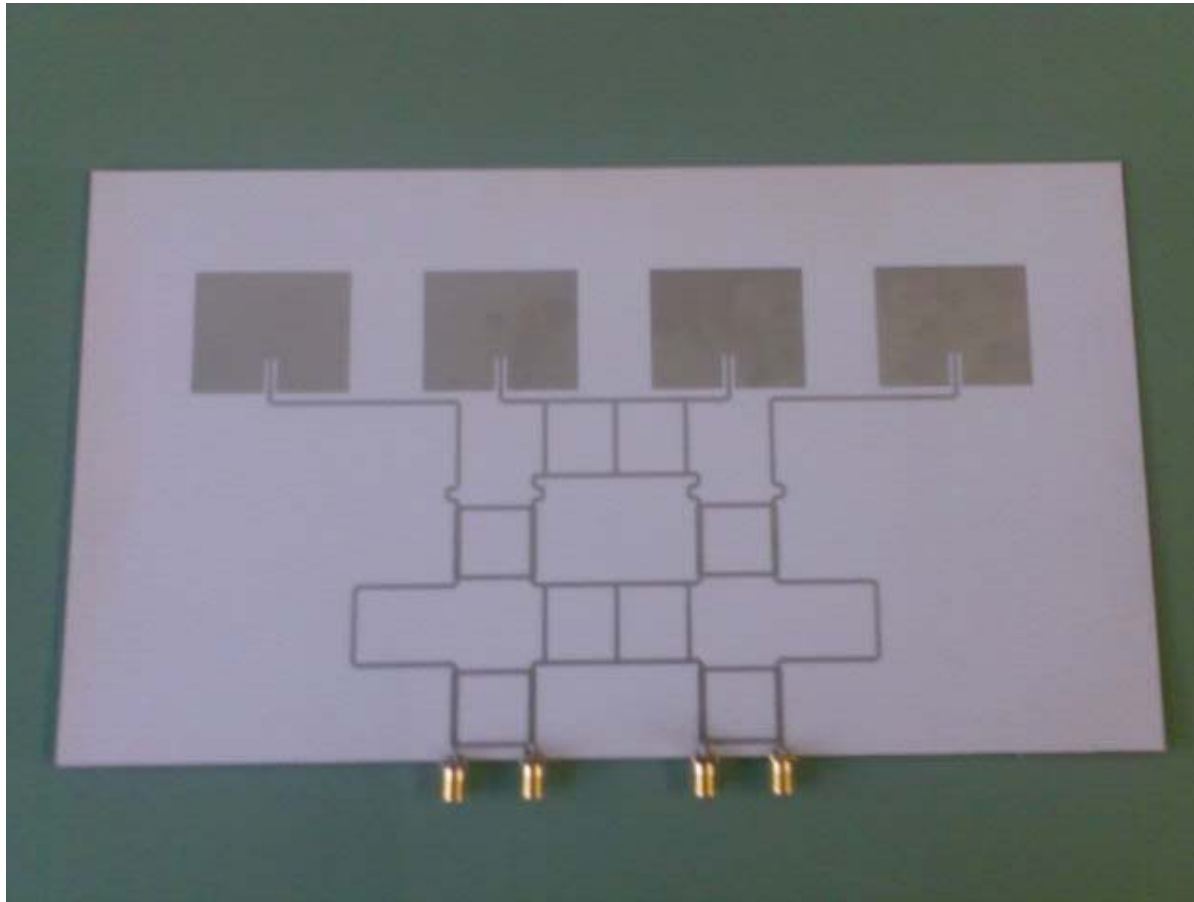
4. Manufacture and test of the Butler Matrix

4.1 Manufacture of the Butler Matrix

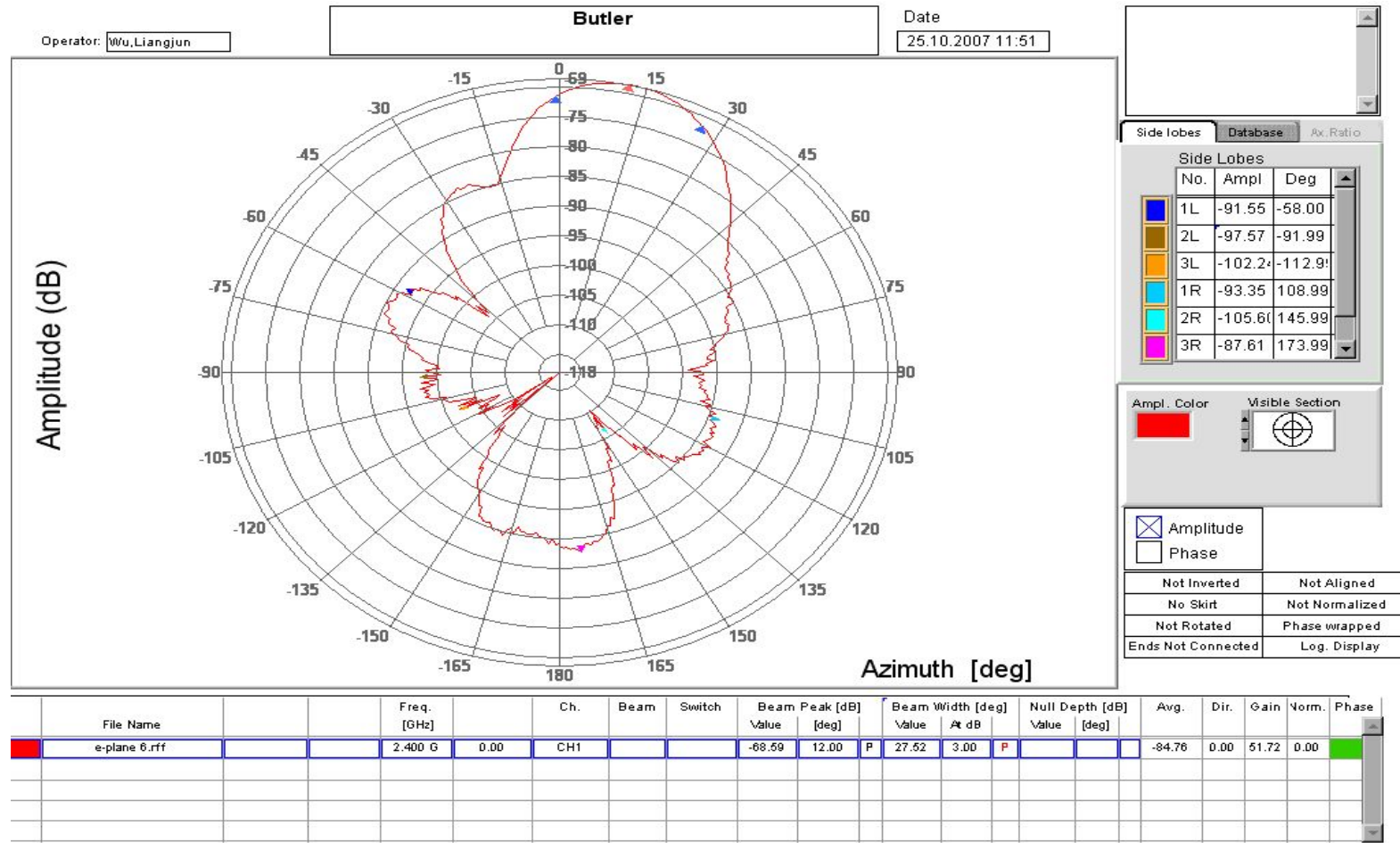
We define the cutline of the Butler-Matrix, then export this design as Gerber file, send the file cond.gbr to the factory to produce the Board.



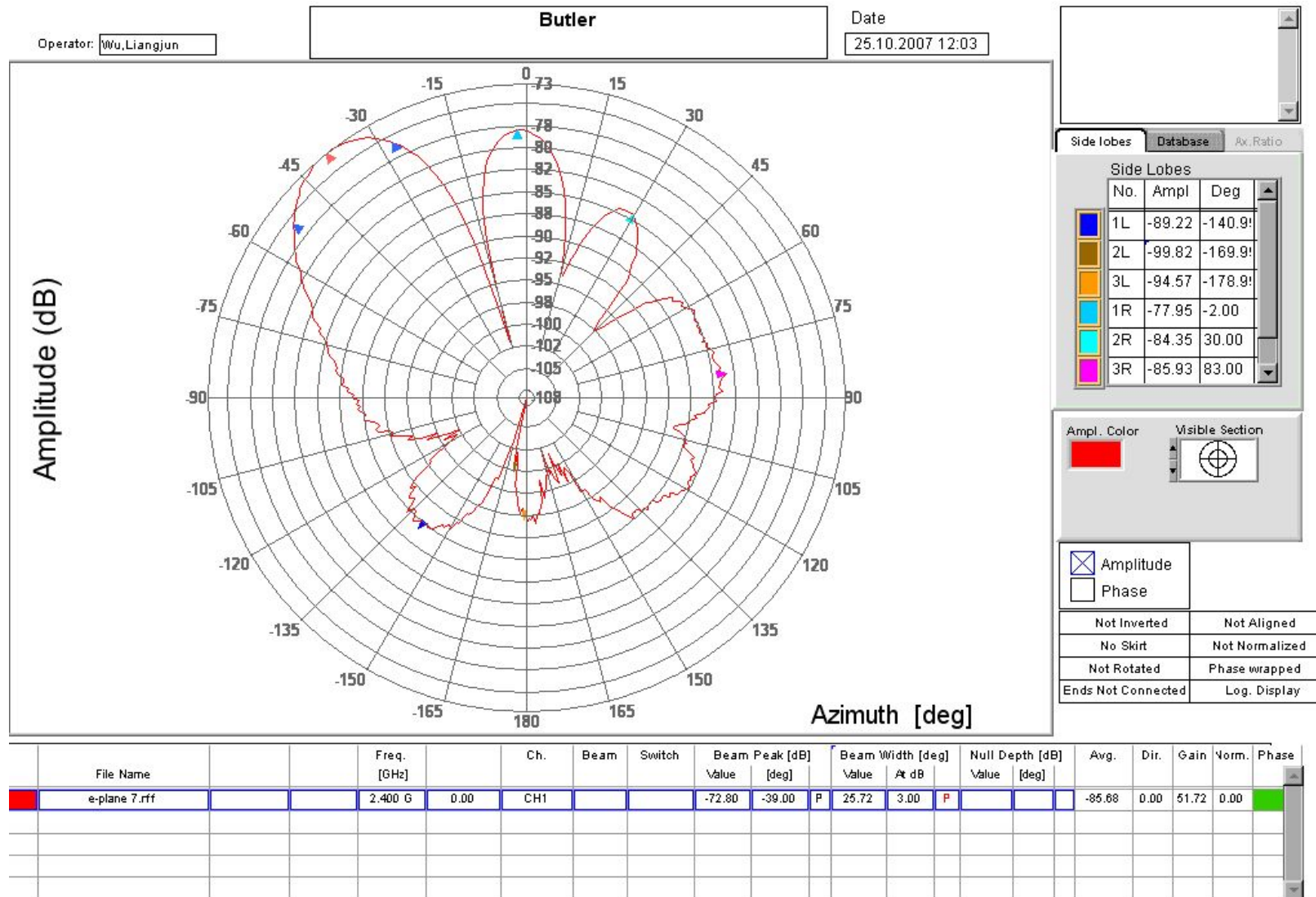
Then we have the finished Butler-Matrix with 4 coupled devices



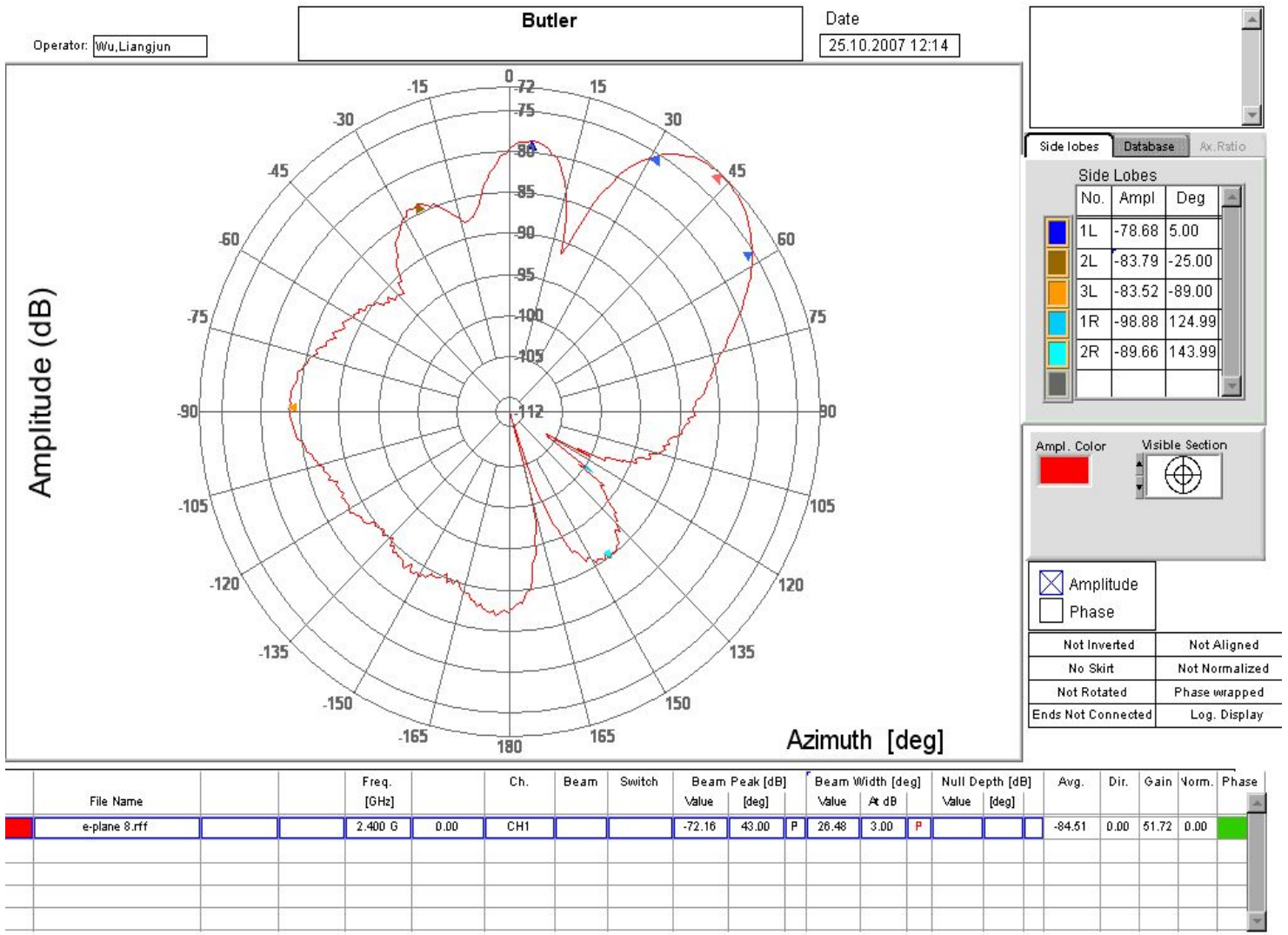
4.2 Test results of the Butler Matrix



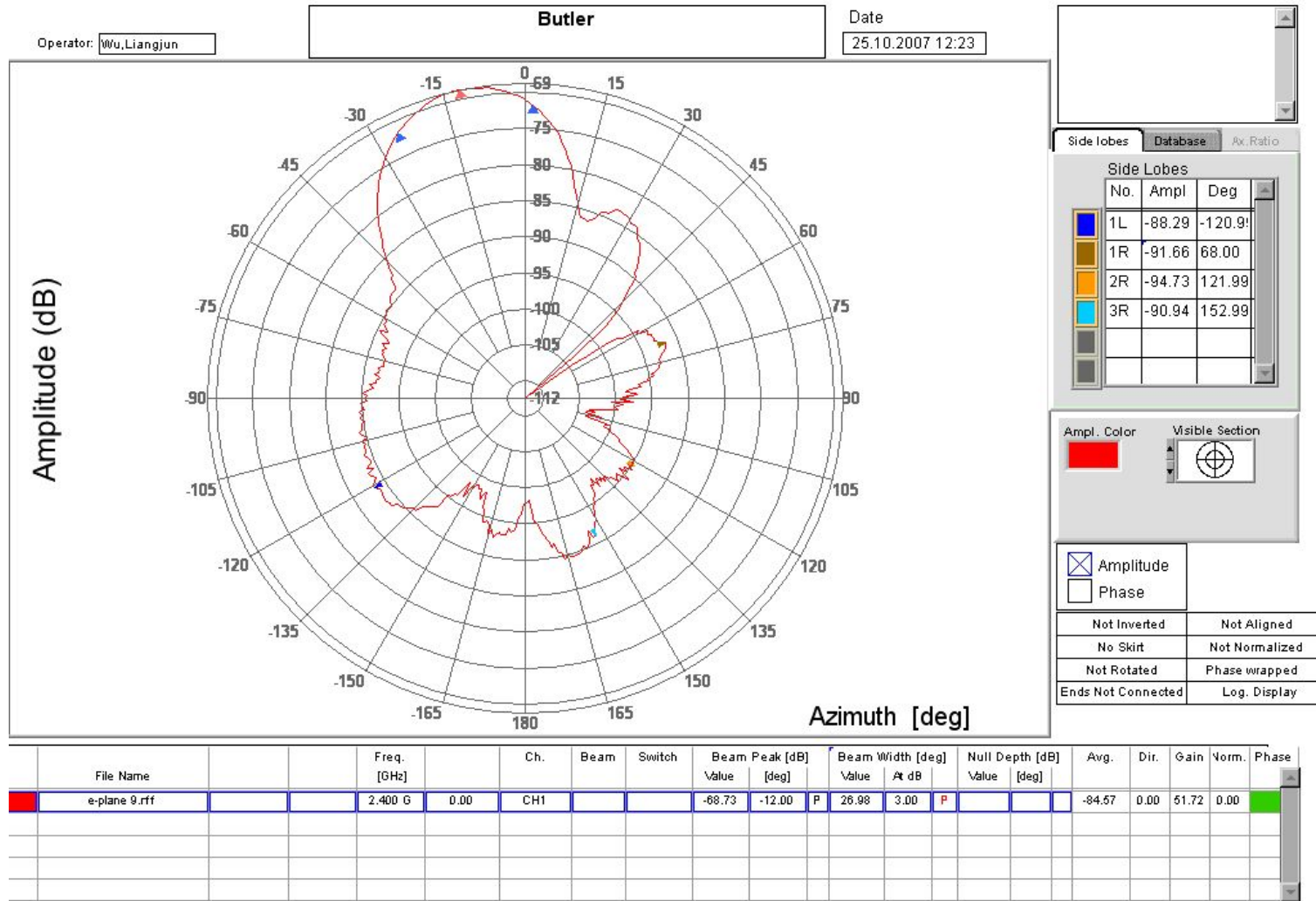
Antenna diagram of Butler Matrix for input P1 (Polar)



Antenna diagram of Butler Matrix for input P2 (Polar)



Antenna diagram of Butler Matrix for input P3 (Polar)



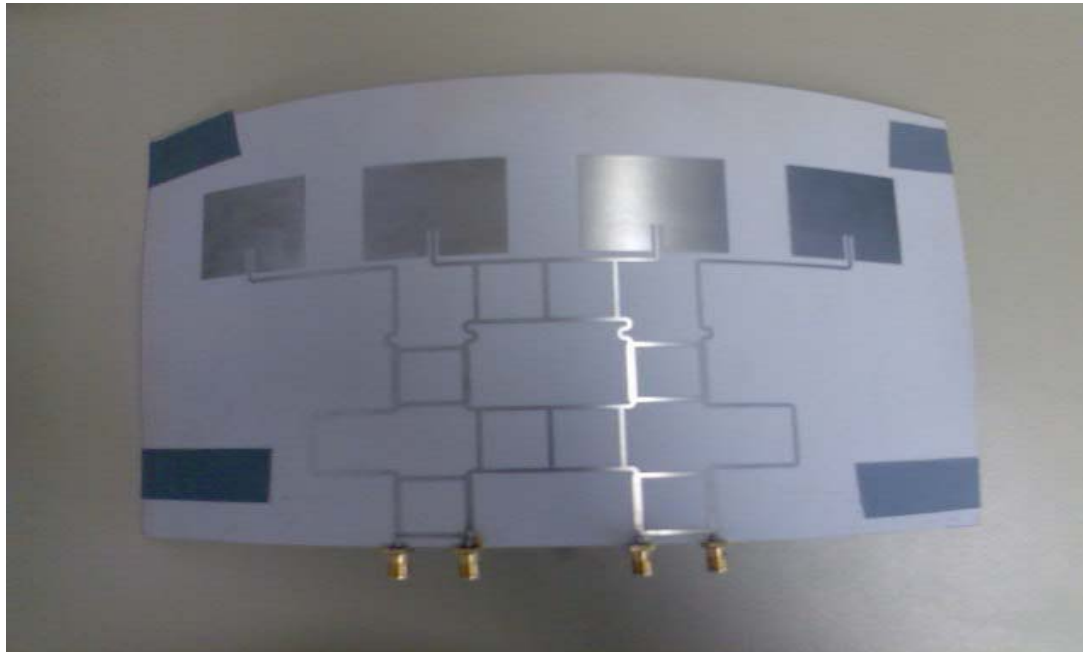
Antenna diagram of Butler Matrix for input P4 (Polar)

We have

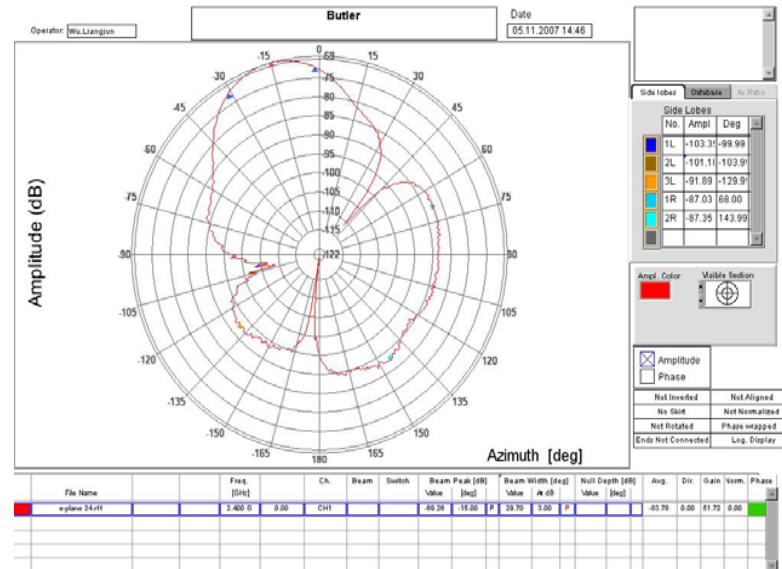
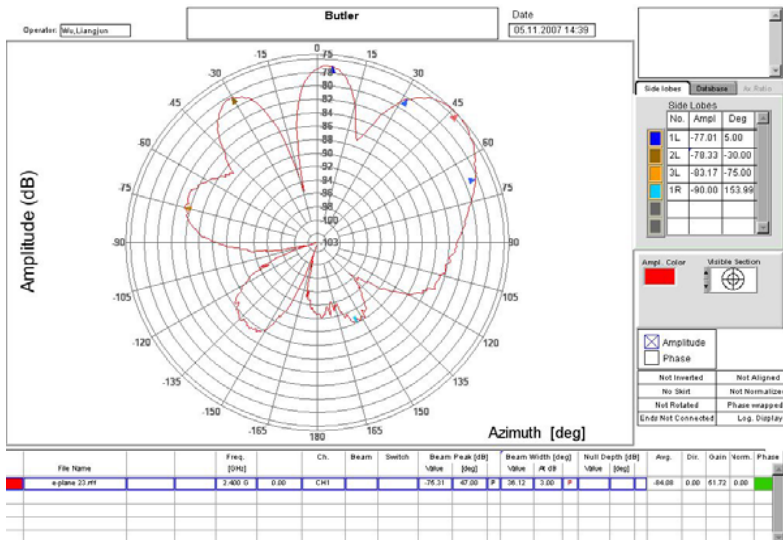
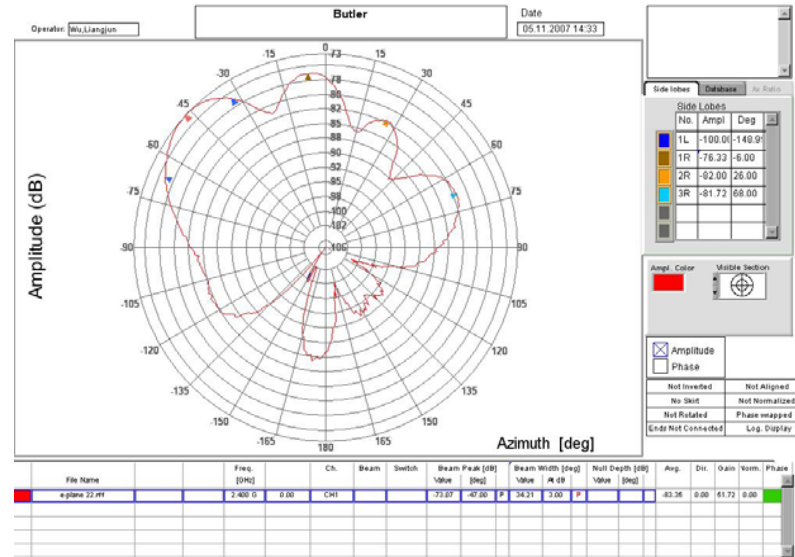
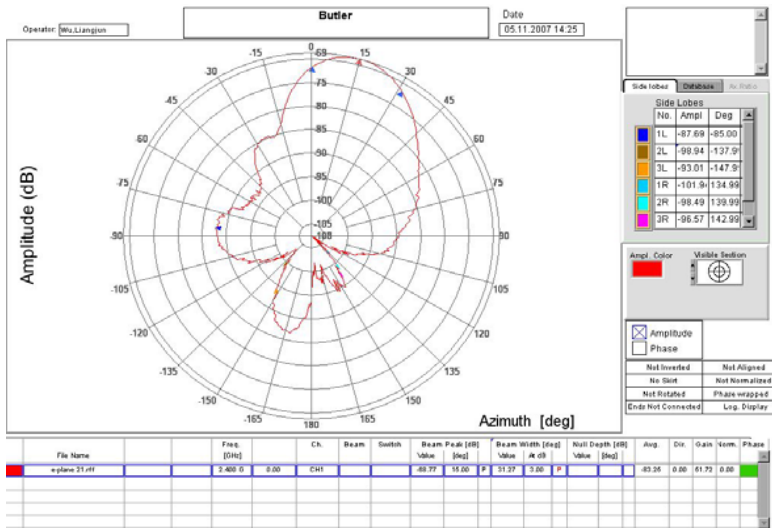
	(theoretic) θ_i	(simulation) θ_i	(test) θ_i
P1	14.5°	12°	12°
P2	-48.6°	-42°	-39°
P3	48.6°	42°	43°
P4	-14.5°	-12°	-12°

4.3 Test results of the Butler Matrix with bend

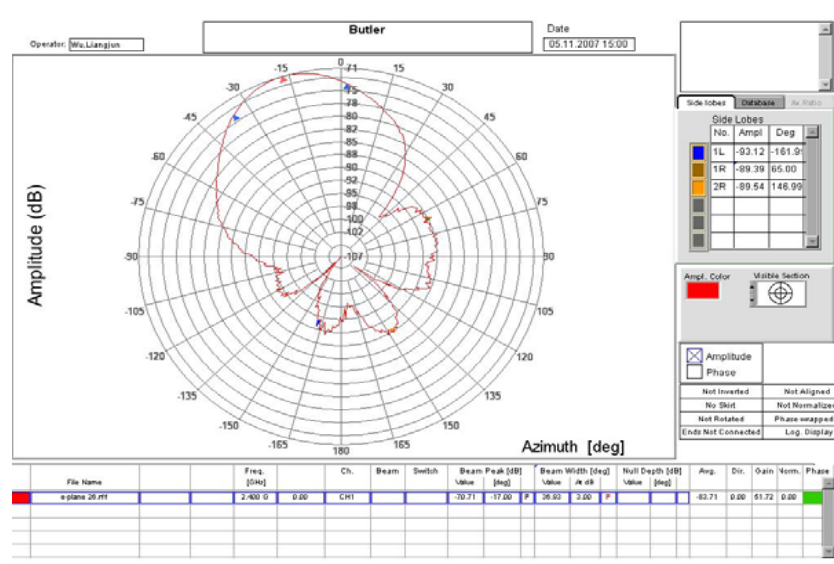
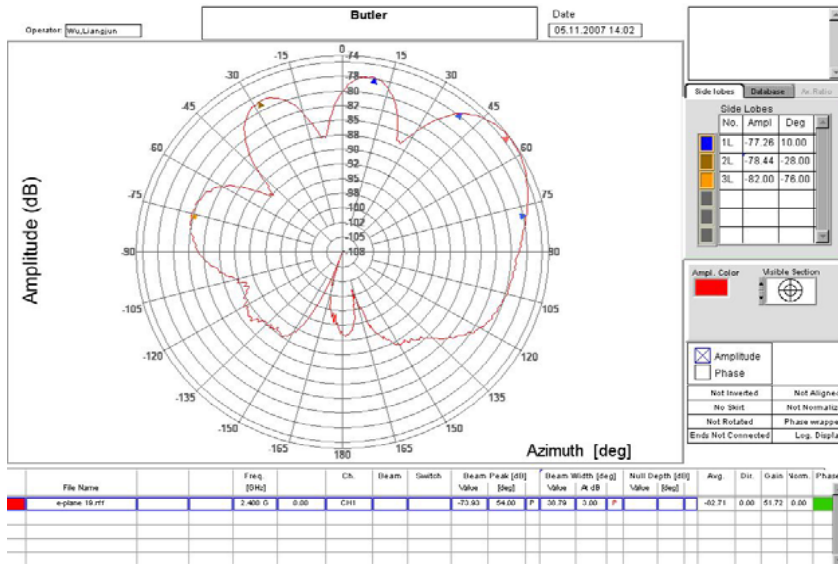
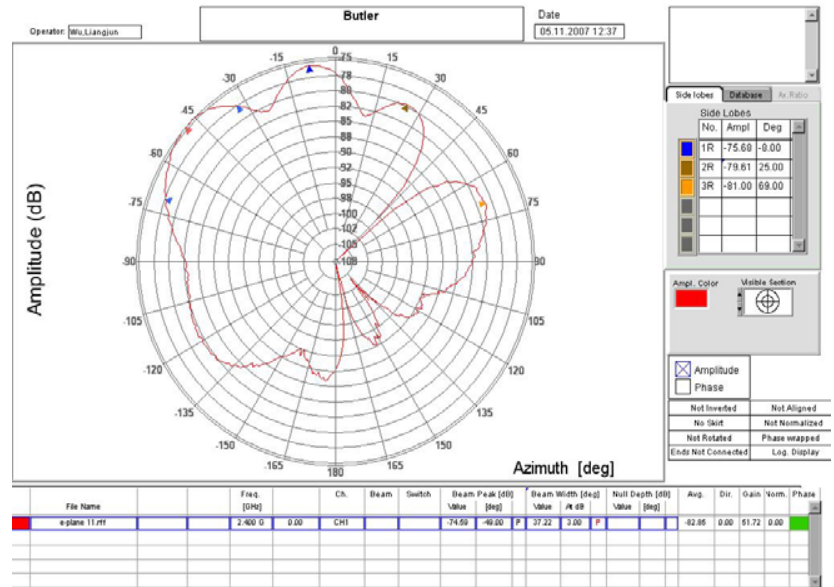
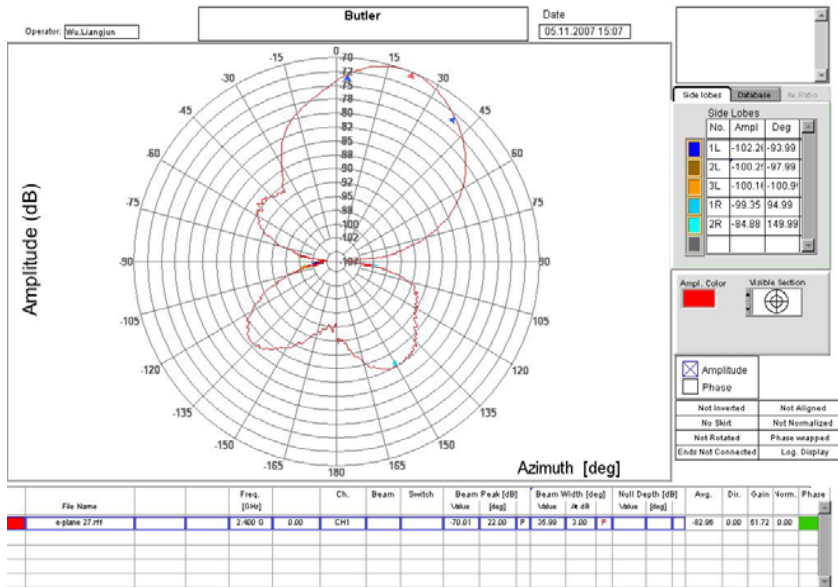
Now we can experimentally bend the board



For L=30mm, the test results are shown in Figure



And for L=50mm, the test results are shown in Figure



We have

	Without bending		With bending (L=30 mm)		With bending (L=50 mm)	
	θ_i	Beam Width	θ_i	Beam Width	θ_i	Beam Width
P1	12	27.57	15	31.27	22	35.99
P2	-39	25.72	-47	34.21	-49	37.22
P3	43	26.48	47	36.12	54	38.79
P4	-12	26.98	-15	29.70	-17	36.93

Analysis of the test results (degree)

From the Table, we know that, when we increase the bend L, we can get bigger and bigger Beam Width. By bending of the board, we can get a bigger coverage area.

5. Conclusion

In this thesis we have designed the 4×4 Butler Matrix (at frequency 2.4 GHz), which consists of four 3dB Hybrid-Coupler, two Cross-Coupler, and four Patch-Antennas.

The test result of the Butler Matrix is just like expected. With four different inputs we can switch the direction of the main beam. (Switched-Beam antenna).

By reducing the element spacing and by suitable bending of the planar substrate carrying the radiator elements, we can get coverage of a full 180° angular in azimuth (hemisphere).

Thank you for your attention!