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## **Master Thesis**

**Mobile Phone Antenna Modelling** 

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## Contents

- Introduction
- Theoretical Background
- Antenna Measurements on Different PCB Variations
- Investigation of the Chassis
- Investigation of Coupling between the Antenna Element and Chassis
- Modelling and Optimizing the Structure (Antenna Chassis Combination)
- Conclusion

## Introduction

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### Currents excited on a Mobile



- Integrated Antenna element couples strongly with the PCB
- PCB acts as a second radiator
  - More Efficiency
  - More Antenna Bandwidth



## Introduction

#### The thesis task

- Design and test of Monopole and PIFA antennas on a large ground plane
- Reflection coefficient measurement of antennas mounted on PCB at
  - Different Positions
  - Orientations
  - PCB Length Variations
- Test of the PCB as a dipole antenna via a balun and coaxial cable
- Investigation of the coupling between the antenna element and PCB by 2 port measurements at different antenna element
  - Positions
  - Orientations
- Developing an equivalent circuit representation of the antenna element, the coupled PCB and the coupling mechanism



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### **Potential Bandwidth:**

- The bandwidth for the best matching case
- The impedance of the center frequency is selected as

the characteristic impedance.

- The reflection coefficient is approximately zero for the center frequency
- So, the potential bandwidth could be found between the -10 dB frequency points.



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Antenna Designs on Large Ground Plane (42.5 cm X 50 cm)

- No Simulation Tool
- Monopole Antenna
  - Center Frequency: 1.047 GHz •
  - Bandwidth: 12.3 % •
  - Gain Maximum at 1 GHz: 2.261 Db

### **PIFA Antenna**

- Center Frequency: 1.048 GHz •
- Bandwidth: 7.86 % •
- Gain Maximum at 1 GHz: 2.778 Db

### **C-Patch Antenna**

- 1.066 GHz Center Frequency: • 3.48 %
- Bandwidth: •
- Gain Maximum at 1 GHz: 2.316 Db

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### Antenna Position Change on the PCB

- Antennas at Orientation Number 3
- The dots indicate the Feed Points
- $\Delta F = f_{position} f_{large ground}$





### Variation of PCB Length

- Antennas at Orientation Number 3
- The dots indicate the Feed Points
- PCB is cut in every 30 mm between 300 mm to 90 mm





Page 9



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#### **Most Bandwidth**

- Orientation Number 3
- Antenna located at the top edge of the PCB
- Around 130 mm PCB Length

### PIFA Antenna



### C-Patch Antenna



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- The chassis is cut into two arms and measured as a dipole antenna
  - 2 baluns and 4 different methods
- Providing Symmetric excitation using an unbalanced coaxial feeding cable





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## First, measuring the chassis as a monopole

- Monopole is conversion of Dipole Antenna
- The possibility to use the unbalanced coaxial feed
- 15 degree dipole antenna measurement in the frequency range of 80-2000 MHz
- Resonance Impedance 19 ohm, so for dipole, it should be around 38 ohm
- Resonance frequency 646 MHz







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- Finally, 1:1 RF Transformer
  Method (350-1500 MHz)
- The tests of open, short terminals and 10, 56, 220 ohm resistors are done.
- Finally, the reference plane is selected as primary port of the transformer and the balun is excluded by the formula;

$$\Gamma_{in} = S_{11} + \frac{S_{21}S_{12}\Gamma_L}{1 - S_{22}\Gamma_L}$$

- Γ<sub>in</sub> is the reflection coefficient of the dipole
- Improved by zeroizing  $s_{11}$  and  $s_{22}$ and by using a ferrite core to suppress the current flow on the outer part of the coaxial cable



- The reason of the spreads in the 10, 56, 220 ohm tests is analyzed as non-linear phase variation with frequency of the RF Transformer which is not correctly compensated by the electrical delay.
- 15 degree of non-linear insertion phase
- In the simulation of 30 degree dipole;
  - Impedance value changes between 34.5 to 121.7 ohm
  - Average impedance is 84 ohm
- Very high error percentage at 10 ohm
- More accurate than 10 ohm at 220
- Acceptable error percentage closer to 56 ohm
- 350-1200 MHz interval is found as acceptable







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- Reflection Coefficient of 30 degree dipole antenna measured in the frequency range of 350-1200 MHz
- The center frequency is 747 MHz and the resonance impedance is 27 ohm
- PBW is 42 %
- Comparable with the simulations







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- The gain of the 30 degree dipole antenna measured at GHz
- E-Plane
  - 5 Db maximum gain
  - 70 degree HPBW
  - Dynamic range 32 Db
- H-Plane
  - 6.44 Db maximum gain
  - 87 degree HPBW
  - Dynamic range 6.5 Db
- Difference is because of the error in the measurement construction



E-Plane

H-Plane



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- Coupling between 30 degree
  Chassis and 1 GHz Antennas
- 2 port measurements to measure  $S_{12}$ and  $S_{21}$
- RF Balun externalization

$$S_{12 \ Coupling} = \frac{S_{12 \ Measured}}{S_{12 \ Transform \ er}}$$

• Antenna elements located at 5 different positions at 3 different orientations on the dipole antenna



Position 1 and 2

Position 3 and 4

Position 5



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- The bandwidth of the antenna-chassis system is related to the amount of coupling
- A part of the antenna element is not over the dipole at the first position
- At position 2, it is totally over the dipole, this is because position 2 has more coupling
- Two radiators at 747 and 1000 MHz frequencies

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The coupling of 2 antennas at the same frequency

- 1. Realization at 747 MHz
- Monopole antenna at 747 MHz is tested on 747 MHz chassis
- The coupling is more if they are at the same frequency as expected. So, more BW is achieved







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- 2-port Simulations of Monopole-Dipole and Monopole-PIFA structures at position 2
- Frequency range of 350-1200 MHz
- Very close results
- Simulation approves the measurement





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## **Modeling and Optimizing the Structure**

- Monopole Antenna Model
  - Series characterization at the first resonance
  - Resonance impedance is 34.5 ohm
  - $\omega = \sqrt{\frac{1}{LC}}$  corresponds to
    - L1 C1 = 25.33 nH pF where the frequency is 1 GHz
  - A parallel capacity to match the measured results better





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- Improved Chassis Model (at 1.09 GHz)
- 30 degree dipole antenna
- Series characterization at the first resonance
- Same resistor value
- Series Capacitance and Inductance are tuned proportionally





### **Modeling of Coupling**

- Two modeling parameters for the coupling;
  - Matching to the Transmission Parameters of the Measurements
  - 1-Port measurement from the antenna side when the dipole is shorted (chassis) must give more BW than the antenna alone

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Antenna Positions at position 2 are used

**Monopole-Dipole Improved Model** 



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1-port measurement from the monopole side when the dipole (chassis) is shorted;



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**PIFA-Dipole Improved Model** 

2-port coupling results;





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• 1-port measurement from the PIFA side when the dipole (chassis) is shorted;



### **Comparison of the Model and Realization**

	Antenna Element Alone	Antenna- Chassis Model	Antenna- Chassis Realization
Monopole	12.3 %	16 %	23 %
PIFA	7.86 %	29 %	25 %
C-Patch	3.48 %	17 %	10 %



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### Conclusion

- The antennas are tested on large and small ground plane variations
- The chassis is investigated as a dipole
- The couplings between the antenna element and chassis are measured. It is seen that the **bandwidth** is related to the **coupling**.
- The equivalent networks of antenna elements, coupled chassis and the coupling mechanism in between are designed
- The chassis and coupling equivalents are improved after tuning the chassis center frequency to 1.09 GHz



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# Thank you for your attention



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